

# Impact of Intrinsic Channel Scaling on InGaAs Quantum-Well MOSFETs

Jianqiang Lin, *Student Member, IEEE*, Dimitri A. Antoniadis, *Life Fellow, IEEE*,  
and Jesús A. del Alamo, *Fellow, IEEE*

**Abstract**—Using a novel gate-last process scheme that affords precise channel thickness control, we have fabricated self-aligned InGaAs quantum-well (QW) MOSFETs. Devices with a channel thickness between 3 and 12 nm, and a gate length between 40 nm and 5  $\mu\text{m}$  are fabricated on a heterostructure that includes a composite InGaAs/InAs QW and an InP barrier. It is observed that channel thickness has a strong impact on the device characteristics. In general, a thick channel is beneficial to ON-state figures of merit, including transconductance and effective carrier mobility. However, a thin channel is beneficial to OFF-state metrics, such as subthreshold swing and drain-induced barrier lowering (DIBL). The InAs composition and effective mass that electrons experience in the channel emerges as a factor that significantly affects channel mobility and presumably the transport characteristics of these devices. The subthreshold swing and DIBL are found to follow a classic scaling behavior. This suggests that the InGaAs QW MOSFETs are at the limit of scaling around  $L_g = 50$  nm.

**Index Terms**—III-V, MOSFETs, quantum-well (QW), self-aligned.

## I. INTRODUCTION

InGaAs is a promising candidate as an n-type channel material for future CMOS due to its superior electron transport properties [1], [2]. Great progress has recently taken place in demonstrating InGaAs MOSFETs. Several self-aligned device architectures have been demonstrated, such as recessed gate [3], [4], implanted source and drain (SD) [5], [6], regrown SD [7], [8], and metallic SD [9], [10]. Among the self-aligned InGaAs MOSFET architectures, the recessed-gate design is an attractive option due to its scalability and simplicity. Recessed-gate InGaAs MOSFETs have delivered the smallest SD contact spacing (footprint) and the smallest contact size of 40 nm in a tight-pitch MOSFET array prototype [11].

High-performance recessed-gate InGaAs MOSFETs are characterized by low parasitics and a simple channel geometry. This has enabled the fundamental device physics studies of

relevance for future logic. For example, we have recently identified the band-to-band tunneling amplified by a parasitic bipolar effect as the cause of excess OFF-state leakage current at high drain bias in InGaAs MOSFETs [12], [13]. Mitigating this is an important priority for future scaled InGaAs MOSFETs [7], [14], [15].

In this quest, there has been increasing concern about transport characteristics being degraded by an aggressive structural design directed to mitigate short-channel effects (SCEs). Split- $C-V$  measurements on long-channel devices have shown that the ultrathin channels suffer from significant mobility degradation [16]. This can be the result of increased scattering and a heavier carrier effective mass from a thinner channel [16]. These mechanisms will also affect short-channel transistors with gate lengths comparable with the mean-free-path, or close to the ballistic limit. However, to date, there have been few experimental studies on the high-performance InGaAs MOSFETs of different channel thicknesses with gate lengths that span from the drift-diffusion regime to the ballistic transport regime, and address both carrier transport and SCE control. Our device technology demonstrated in [11] allows us to tightly control the channel thickness through a precision recess method. At the same time, the gate length of functional devices can be scaled down to 40 nm. This device technology then represents an ideal platform to study the tradeoffs involved in a channel thickness design as the gate length is scaled.

This paper is organized as follows. In Section II, we briefly describe the key features of our transistor and its fabrication process, particularly the engineering scheme that we use to achieve precise channel thickness control. In Sections III and IV, we show the impact of intrinsic scaling on ON-state and OFF-state characteristics. This includes both the channel thickness (vertical) scaling and the gate length (lateral) scaling. We finally state our key conclusions in Section V. This paper refines and extends the experimental study in [11].

## II. DEVICE FABRICATION

A cross-sectional schematic of a typical device used for this paper is shown in Fig. 1. This schematic shows a surface-channel design, in which the  $n^+$  cap structure has been recessed to expose the quantum-well (QW) channel. A buried-channel design is also possible if the etching does not completely remove the thin InP barrier (also behaving as etch-stop) above the as-grown channel.

Manuscript received April 6, 2015; revised May 25, 2015; accepted June 3, 2015. Date of publication July 31, 2015; date of current version October 20, 2015. This work was supported in part by the Defense Threat Reduction Agency under Contract HDTRA 1-14-1-0057, in part by the National Science Foundation through the Center for Energy Efficient Electronics Science—Science and Technology Center under Award 0939514, and in part by the Singapore-MIT Alliance for Research and Technology Centre. The review of this paper was arranged by Editor J. S. Suehle.

The authors are with Microsystems Technology Laboratories, Massachusetts Institute of Technology, Cambridge, MA 02139 USA (e-mail: linjq@mit.edu; daa@mit.edu; alamo@mit.edu).

Color versions of one or more of the figures in this paper are available online at <http://ieeexplore.ieee.org>.

Digital Object Identifier 10.1109/TED.2015.2444835

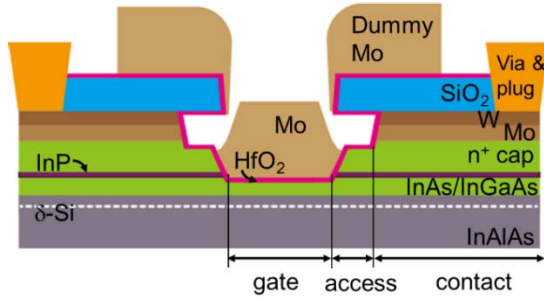


Fig. 1. Cross-sectional schematic of a self-aligned surface-channel InGaAs QW-MOSFET fabricated for this paper. Buried-channel designs are also possible by leaving the InP barrier above the channel in the intrinsic gate portion of the device.

The devices are fabricated by a self-aligned process as reported in [11] and in more detail in [17]. A simplified description is given here. The heterostructure is grown by molecular beam epitaxy on an InP substrate and it comprises from bottom to top of a  $\sim 400$ -nm-thick InAlAs buffer with a Si  $\delta$ -doped sheet placed 5 nm beneath the channel, an undoped  $\text{In}_{0.7}\text{Ga}_{0.3}\text{As}/\text{InAs}/\text{In}_{0.7}\text{Ga}_{0.3}\text{As}$  channel with a layer thickness of 5/2/3 nm, a 3-nm undoped InP barrier, and a composite  $n^+$  cap containing  $n^+$  InAlAs,  $n^+$  InP, and  $n^+$   $\text{In}_{0.7}\text{Ga}_{0.3}\text{As}$ . The Si doping level in the cap is between  $2 \times 10^{19}$  and  $3 \times 10^{19} \text{ cm}^{-3}$ . Unless indicated, all layers are lattice matched to the InP substrate.

Device fabrication begins with a blanket deposition of 15-nm sputtered Mo, 15-nm sputtered W, and 50-nm CVD  $\text{SiO}_2$ . E-beam lithography is used to define the gate region.  $\text{SiO}_2$ , W, and Mo are then sequentially etched using F-based reactive ion etching (RIE), which also requires a proper annealing to repair the RIE damage [18]. In this process, an overhang is created that defines the length of the access region.

The III–V cap recess is then performed. The III–V layers are first recessed by a Cl-based RIE [15], followed by digital etch. The III–V RIE is self-aligned to the  $\text{SiO}_2$  hard mask, and stops a few nanometers above the InP barrier. The digital etch is made of two steps: 1) oxide formation by low power  $\text{O}_2$  plasma and 2) oxide stripping by dilute sulfuric acid. When separated, both steps are self-limiting and, under our selected conditions, they remove  $\sim 1$  nm of semiconductor materials per full cycle [19].

After the last cycle of digital etch, a fresh semiconductor surface is exposed, immediately after which, 2.5-nm  $\text{HfO}_2$  is deposited as a gate insulator by atomic layer deposition (ALD), followed by Mo gate electrode evaporation. The device is completed with gate patterning, via opening, and metal pad deposition.

Accurate channel thickness control is the heart of this study. To realize this, during the digital etch, we use a procedure that allows us to pinpoint the precise location of the top surface of the InP etch-stop layer [17]. From that point on, we can calculate the number of digital etch cycles needed to achieve a target channel thickness. Here, we use  $t_c$  to denote the total channel thickness left after the III–V recess, that is, the total semiconductor thickness above the InAlAs buffer in the intrinsic

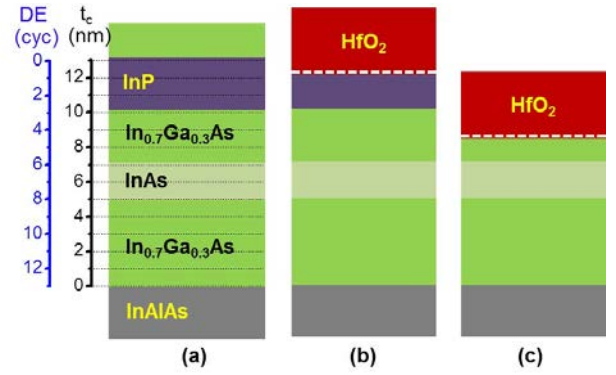


Fig. 2. (a) Correlation between the number of digital etch cycles and the final channel thickness. Examples of (b) buried channel and (c) surface channel designs. Dashed lines beneath the  $\text{HfO}_2$ : location of the oxide/semiconductor interface.

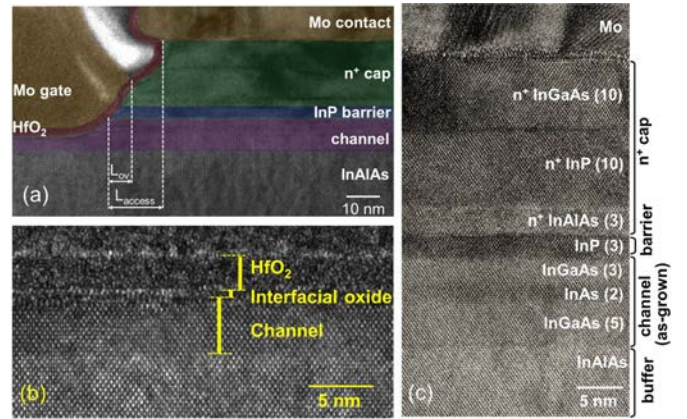


Fig. 3. (a) False-color TEM cross section of a finished QW-MOSFET around the edge of the gate. The final channel thickness of this device is 4 nm. The access region ( $L_{\text{access}}$ ) and gate- $n^+$  overlapping region ( $L_{\text{ov}}$ ) are indicated. (b) HRTEM cross section of the intrinsic gate region. (c) HRTEM of the contact region. Bracket after layer label indicates thickness in nanometers.

region of the device. The correlation between the number of digital etch cycles and the final channel thickness, and the examples of buried channel and surface channel designs are shown in Fig. 2(a)–(c). For  $t_c$  between 11 and 13 nm, the high- $\kappa$  dielectric interface is located on the InP barrier, resulting in a buried channel device [Fig. 2(b)]. For  $t_c$  of 10 nm or below, the high- $\kappa$  dielectric directly lies on the QW channel, resulting in a surface channel device [Fig. 2(c)]. For  $t_c$  below 5 nm, the high-mobility InAs core in the channel is entirely removed. The estimated  $t_c$  involves 1-nm uncertainty due to the discrete nature of our digital etching technique [19].

Fig. 3(a) shows a cross-sectional transmission electron microscopy (TEM) image of the transition region between the intrinsic channel on the left and the extrinsic region on the right in a finished device with  $t_c = 4$  nm. False color has been added to highlight the different layers. The access region is defined as the spacing ( $L_{\text{access}}$ ) between the edge of the intrinsic channel and the edge of the contact. For this device,  $L_{\text{access}}$  is 15 nm. Within this small access region, a gate- $n^+$  overlap region ( $L_{\text{ov}}$ ) is created by introducing a slight angle and sample rotation during gate evaporation.  $L_{\text{ov}}$  spans a lateral dimension of  $\sim 7$  nm. The actual  $n^+$ -gate

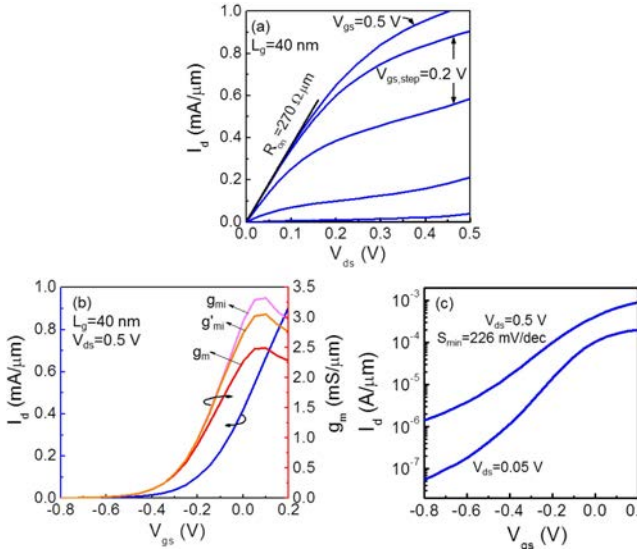


Fig. 4. Electrical characteristics of device with  $t_c = 9$  nm and  $L_g = 40$  nm. (a) Output characteristics. (b) Transfer and transconductance characteristics at  $V_{ds} = 0.5$  V. (c) Subthreshold characteristics. In (b), extrinsic transconductance is labeled as  $g_m$ , intrinsic transconductance based on first-order model as  $g_{mi}$ , and intrinsic transconductance with output conductance correction as  $g'_{mi}$ .

overlap distance is slightly longer due to the side-wall angle. This design with a short access region that partially overlaps with the gate saves device footprint, reduces parasitic series resistance, and maintains sufficient isolation between the gate and the SD contacts.

Fig. 3(b) and (c) are the high-resolution TEM (HRTEM) images for the intrinsic gate region and the extrinsic contact region, respectively. Fig. 3(b) shows that a uniform channel and a flat surface are obtained after the III–V recess. The channel is between 7 and 8 atomic monolayers thick. This agrees with our targeted channel thickness for this device of 4 nm and showcases the excellent accuracy of our process. The 2.5-nm  $\text{HfO}_2$  gate dielectric is also highlighted. Between the dielectric and the channel, an interfacial oxide of 0.5–1 nm is observed. This is probably formed during the transition between the last native oxide stripping and the ALD. It has been reported that ALD- $\text{HfO}_2$  using TEMAH precursor, as is the case here, results in a self-cleaning effect on the InGaAs surface. This refers to the partial removal of the native oxide, which mitigates Fermi-level pinning at the  $\text{HfO}_2/\text{InGaAs}$  interface [20]. In our case, some residual native oxide appears to remain at the InGaAs interface. Electron trapping defects in this interfacial layer might result in  $V_t$  instability after the prolonged gate stress [21]. Fig. 3(c) also shows the Mo ohmic contact on InGaAs. A smooth, unalloyed contact interface free of any interfacial oxide is evident.

### III. ON-STATE CHARACTERISTICS

Electrical characteristics of a short-channel transistor are shown in Fig. 4. This device has  $t_c = 9$  nm (surface-channel) and  $L_g = 40$  nm. The output characteristics in Fig. 4(a) show an excellent ON-resistance of  $270 \Omega \cdot \mu\text{m}$ . At  $V_{ds} = 0.5$  V, the transconductance and subthreshold characteristics in Fig. 4(b) and (c) show a peak extrinsic transconductance ( $g_{m,\max}$ ) of  $2.5 \text{ mS}/\mu\text{m}$  and a minimum subthreshold

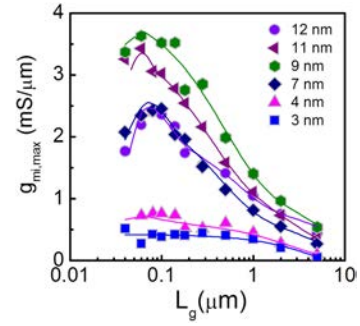


Fig. 5. Peak intrinsic transconductance at  $V_{ds} = 0.5$  V versus  $L_g$  for different  $t_c$ .

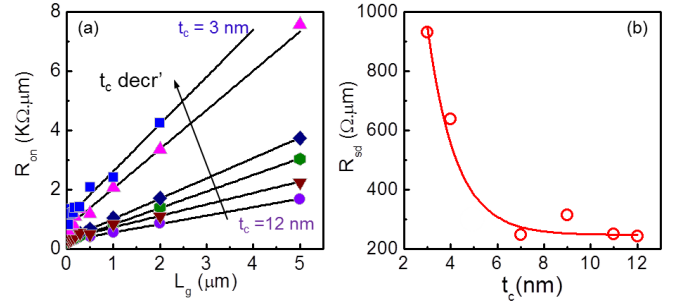


Fig. 6. (a)  $R_{ON}$  versus  $L_g$  for  $t_c$  from 3 to 12 nm, in which  $R_{sd}$  is extracted from the y-intersection. (b)  $R_{sd}$  versus  $t_c$ .

swing ( $S_{\min}$ ) of 226 mV/decade. Due to the relaxed channel thickness and the extremely short channel of this device, a significant SCE is observed. A systematic study on SCE will be discussed later. In the same chip, a hero device with  $L_g = 80$  nm delivers a record  $g_{m,\max}$  of  $3.1 \text{ mS}/\mu\text{m}$  and  $R_{ON}$  of  $190 \Omega \cdot \mu\text{m}$  [11].

We select the intrinsic peak transconductance ( $g_{mi,\max}$ ) as the metric to study the scaling behavior of ON-state performance. The intrinsic transconductance ( $g_{mi}$ ) is estimated using the method in [22]. Here it does not take into account the effect of the output conductance because the worst case (highest output conductance) of the shortest gate length device with the thickest channel, this correction resulted in a relative small error (8%) in the estimation of  $g_{mi,\max}$  as shown in Fig. 4(b). Fig. 5 graphs  $g_{mi,\max}$  versus  $L_g$  for families of devices with different  $t_c$ . We find that  $g_{mi,\max}$  follows a classic gate length scaling behavior: it increases as  $L_g$  decreases and it eventually saturates. For  $t_c > 7$  nm,  $g_{mi,\max}$  decreases as the gate length further decreases. This is the result of a significant SCE.

In extracting  $g_{mi}$ , we use the measured source resistance  $R_s$ . This is obtained from a measurement of the ON-resistance ( $R_{ON}$ ), which is performed at a high gate overdrive of 0.8 V. Fig. 6(a) shows  $R_{ON}$  as a function of  $L_g$  for  $t_c$  from 12 to 3 nm. For all  $t_c$  values, a linear dependence of  $R_{ON}$  with  $L_g$  is obtained. The intercept of these lines with the y-axis gives the value of  $R_{sd}$ , the sum of the SD resistance. Fig. 6(b) shows  $R_{sd}$  as a function of  $t_c$ .  $t_c > 7$  nm devices exhibit  $R_{sd}$  of  $\sim 250 \Omega \cdot \mu\text{m}$  independent of  $t_c$ . This excellent result is due to our  $n^+$ -gate overlap design and good contact resistance [23]. For thin channels,  $R_{sd}$  rapidly increases as  $t_c$  reaches 4 nm and below. This is presumably due to the increased spreading resistance associated with the link region at the gate edge of



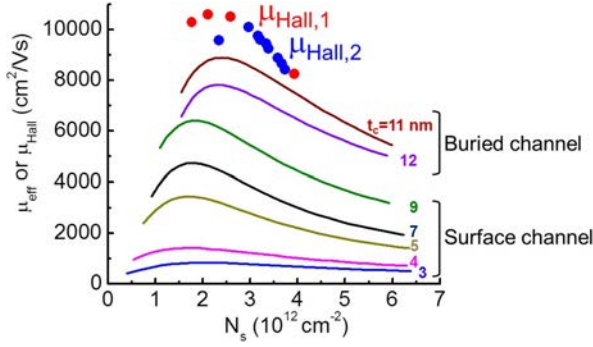


Fig. 7. Mobility versus sheet charge density. Continuous lines: effective channel mobility in long-channel devices ( $L_g = 5 \mu\text{m}$ ) with different channel thicknesses. Solid symbols: two sets of etch-Hall mobility from two growth test structures (see text).

the channel. Further investigation is required to understand this behavior. Due to the symmetry of the device, we assume that  $R_s$  is half of  $R_{sd}$ .

Fig. 5 reveals that the carrier transport in the intrinsic device is greatly influenced by  $t_c$ . The highest  $g_{mi,max}$  is obtained for  $t_c = 9 \text{ nm}$ , which is the thickest surface-channel device that we have fabricated. For the channels thicker than this, the devices are of buried-channel type and  $g_{mi,max}$  decreases due to a loss in gate-channel capacitance [11]. For the channels thinner than this,  $g_{mi,max}$  rapidly decreases. This reveals worsening transport characteristics.

Independent evidence of the key role of transport in the performance of these devices is obtained from the measurements of channel mobility. The effective field mobility,  $\mu_{eff}$ , is extracted by the split- $C-V$  method in long-channel devices ( $L_g = 5 \mu\text{m}$ ) and the results are shown in Fig. 7 as a function of sheet electron concentration ( $N_s$ ). The  $I_d-V_{gs}$  characteristics used in the mobility extraction were corrected for series resistance. This correction is relatively small: in the worst case (highest  $N_s$  at thick  $t_c$ ), it is 14%. The impact of carrier trapping is not considered. It has been reported that a high concentration of border traps can exist in the high- $\kappa$  oxide close to the III-V interface and causes significant trapping especially at high gate voltage [24], [25]. This could lead to error in the mobility extraction at a high carrier concentration.

Fig. 7 also includes the measurements of Hall mobility versus a sheet electron concentration from two epi calibration test structures that were grown immediately before the actual device heterostructures. These test samples have the same channel and barrier design as the device heterostructure but a modified cap that consists of  $n^+$  InAlAs (3 nm),  $n^+$  InP (3 nm), and  $n^+$   $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$  (25 nm) with Si doping between  $2 \times 10^{19}$  and  $3 \times 10^{19} \text{ cm}^{-3}$ . Hall mobility measurements are performed as the  $n^+$  InGaAs portion of the cap is thinned down through selective wet etching. This results in a reduction of  $N_s$  and an increase in mobility as the relative fraction of charge in the high-mobility channel increases. A heterostructure with a high Hall mobility has been found to be essential to obtaining a high effective channel mobility of III-V MOSFET. Our results confirm this observation.

Fig. 7 reveals a nearly progressive deterioration of channel mobility as the channel is thinned down. The highest

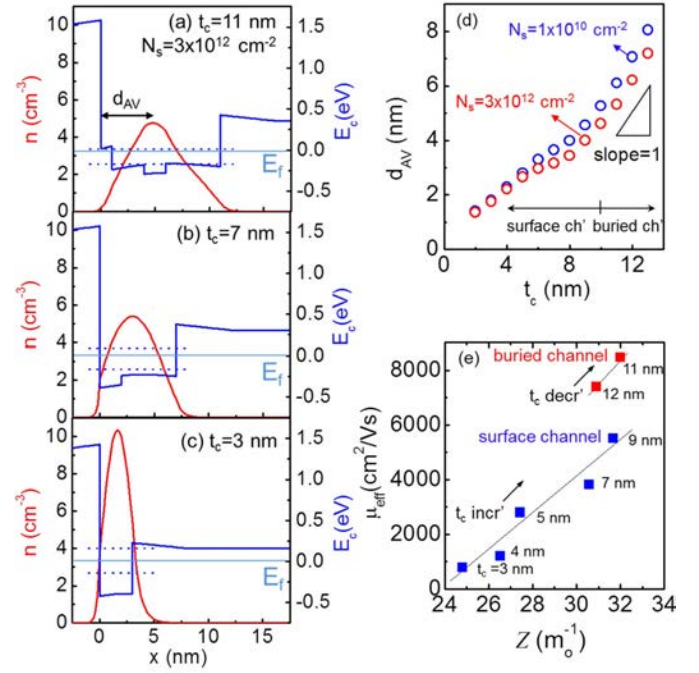


Fig. 8. (a)–(c) Carrier distribution ( $n$ ) and conduction-band energy ( $E_c$ ) in the direction perpendicular to the channel. The semiconductor/oxide interface is placed at  $x = 0$ . The energy of the first two quantum states is shown in dashed lines. (d) Average distance of channel electron population ( $d_{AV}$ ) with respect to the oxide–semiconductor interface at  $N_s$  of  $1 \times 10^{10} \text{ cm}^{-2}$  (blue) and  $3 \times 10^{12} \text{ cm}^{-2}$  (red). The schematic of  $d_{AV}$  is shown in the inset in (a). (e) Experimental mobility versus  $Z$  at  $N_s$  of  $3 \times 10^{12} \text{ cm}^{-2}$  (see text).

peak mobility is obtained in the buried channel device with 1-nm InP barrier where  $\mu_{eff} = 8800 \text{ cm}^2/\text{V} \cdot \text{s}$  at  $N_s$  of  $2.6 \times 10^{12} \text{ cm}^{-2}$ . This is  $\sim 80\%$  of the etch-Hall mobility at the same carrier concentration. With an InP barrier of 2 nm, the mobility is slightly lower. In the surface channel devices, the mobility is lower than in the buried-channel designs and it monotonically drops as the channel becomes thinner. The mobility rapidly falls when  $t_c$  is reduced below 5 nm.

In order to understand these results, we have performed 1-D Poisson–Schrödinger (P–S) simulations (Nextnano<sup>3</sup> [27]) in the intrinsic region of the device. Fig. 8(a)–(c) shows the carrier distribution ( $n$ ) and the conduction-band edge energy ( $E_c$ ) in the direction perpendicular to the channel ( $x$ ) at a sheet electron concentration of  $N_s = 3 \times 10^{12} \text{ cm}^{-2}$  for  $t_c$  of 11, 7, and 3 nm. For a thick channel, the charge in the channel is broadly spread in space and away from the oxide–semiconductor interface. In a very thin channel, the electrons are tightly confined against the interface.

In order to evaluate the role of the charge distribution on transport, we have calculated the distance of the charge centroid in the channel from the oxide–semiconductor interface,  $d_{AV}$ . This is graphically indicated in Fig. 8(a) and is computed using the following expression:

$$d_{AV} = \frac{1}{N_s} \int_0^{t_c} n(x) \cdot x dx. \quad (1)$$

In this equation,  $n(x)$  is the local carrier density at a depth  $x$  from the oxide–semiconductor interface where  $x = 0$ .

Fig. 8(d) graphs  $d_{AV}$  as a function of  $t_c$  at  $N_s$  of  $1 \times 10^{10}$  and  $3 \times 10^{12} \text{ cm}^{-2}$ , which are the scenarios close to threshold and for ON-state, respectively. It is clear that as the channel thickness increases, the charge centroid continuously moves further away from the metal–oxide interface. The trend is more pronounced for buried channels. From these calculations, we can conclude that in thin channel designs, the electrons suffer more from surface roughness scattering as well as remote columbic scattering, both of which would degrade the mobility.

The electrons in the InAs core possess the lowest effective mass and hence the higher mobility. The larger the electron fraction in the InAs subchannel, the higher the overall mobility should be. To quantitatively evaluate this, we extract the weighted average of the inverse of the  $\Gamma$ -valley effective mass ( $m_\Gamma$ ) of the channel electron,  $Z$ , in the unit of  $m_o^{-1}$ . We obtained this from P–S simulations using

$$Z = \sum_i \frac{1}{m_i} \frac{1}{N_s} \int_{t_i} n(x) dx. \quad (2)$$

In this equation, the integrals evaluate the sheet electron concentration in the respective channel layer  $i$ , which can be InP and InAs or  $\text{In}_{0.7}\text{Ga}_{0.3}\text{As}$ .  $m_i$  and  $t_i$  are the effective mass and the thickness of channel layer, respectively. Effective mass of those materials follows the room-temperature values in [28].

Fig. 8(e) graphs experimental channel mobility versus  $Z$ , both at  $N_s$  of  $3 \times 10^{12} \text{ cm}^{-2}$ . The figure reveals a clear correlation: increased  $Z$  results in higher mobility. Among the surface channel devices, the mobility monotonically increases with  $Z$ , which itself increases with  $t_c$ . This reflects the reduced effective mass and weaker alloy scattering associated with a higher relative InAs mole fraction in the channel. At a particular value of  $Z$  ( $\sim 31$ ), there is a step increase in the mobility for the buried-channel designs. This is likely due to the reduction of interface scattering experienced by the electrons due to increased distance of the 2DEG from the oxide interface. As can be observed in Fig. 8(d), the 2DEG is close to the  $\text{HfO}_2/\text{III-V}$  interface for the surface channel devices, and it rapidly moves away from the  $\text{HfO}_2/\text{III-V}$  interface in the buried channel devices. However, in those devices, increased  $t_c$  brings about increased penetration of the electron wave function into the InP barrier, which results in a reduction of  $Z$ . This is the likely reason why now mobility decreases with increasing  $t_c$  from 11 to 12 nm.

The channel thickness dependence of the mobility contributes to explaining the strong impact that a channel thickness has on transconductance (Fig. 5). A complete model for the transconductance requires the development of a charge-control model and a transport model that consider the effective mass under strain in each of the channel layers and include conduction band nonparabolicity. This detailed study is best done on devices with a simpler channel design than those in this paper.

#### IV. OFF-STATE CHARACTERISTICS

Despite the advantages of using a thick channel from a transport point of view, the drawback is evident in poor SCE control. Fig. 9(a) and (b) shows minimum subthreshold

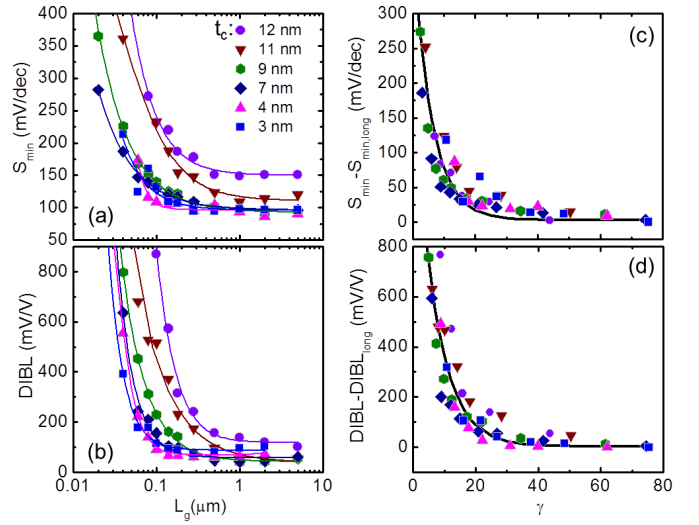


Fig. 9. (a) and (b) Experimental  $S_{\min}$  and DIBL as a function of gate length for  $t_c = 3$  to 12 nm. (c) and (d)  $S_{\min} - S_{\min, \text{long}}$  and  $\text{DIBL} - \text{DIBL}_{\text{long}}$  as a function of  $\gamma$  based on the natural channel length methodology discussed in the text.

swing,  $S_{\min}$ , at  $V_{ds} = 0.5 \text{ V}$  and drain-induced barrier lowering (DIBL) versus  $L_g$  for devices with varying  $t_c$ . DIBL is obtained from  $V_t$  shift between  $V_{ds} = 0.05$  and  $0.5 \text{ V}$  at a constant current of  $I_d = 1 \times 10^{-6} \text{ A}/\mu\text{m}$ . However, not all devices can be turned OFF to this current level; hence some data points at short gate lengths are not available.  $S_{\min}$  and DIBL in Fig. 9(a) and (b) follow a classic scaling behavior; they are both independent of gate length for long  $L_g$  but rapidly degrade as  $L_g$  scales down beyond a certain value. The onset of  $S_{\min}$  and DIBL degradation occurs at longer gate lengths for the devices with thicker  $t_c$ .

These results can be analyzed using classic SCE electrostatics originally developed for fully depleted (FD) SOI MOSFETs in [29]. Electrostatically, the QW-MOSFETs discussed here are similar to FDSOI transistors, where the InAlAs buffer is equivalent to the buried oxide. Based on Yan's model, the natural channel length  $\lambda$  is defined that depends on the geometry and the dielectric constant of the various materials involved

$$\lambda = \sqrt{t_{\text{ins}} t_c \frac{\epsilon_c}{\epsilon_{\text{ins}}}}. \quad (3)$$

In this equation,  $\epsilon_{\text{ins}}$  is the effective dielectric constant of the insulator and  $\epsilon_c$  is that for the channel (both relative values);  $t_c$  is the channel thickness and  $t_{\text{ins}}$  is the effective insulator thickness. Here, we depart from Yan's model: in our case,  $\epsilon_{\text{ins}}$  and  $t_{\text{ins}}$  consider the  $\text{HfO}_2$  gate insulator, the interfacial oxide, and  $d_{AV}$ . For the interfacial oxide, we assume that it has a thickness of 1 nm (Fig. 3) and a relative dielectric constant of 16. The composition and properties of III–V native oxides are not well known. For example, relative dielectric constants between 9.9 and 23.3 have been reported for  $\text{Ga}_2\text{O}_3$  [30], [31]. Here, we assume that the interfacial oxide has a dielectric constant given by the average of these two values. The average distance of channel electron away from the oxide–semiconductor interface,  $d_{AV}$ , arises from the quantum nature of the channel and the presence of the

InP barrier in the buried-channel designs. Here, we use  $d_{AV}$  at  $N_s = 1 \times 10^{10} \text{ cm}^{-2}$  as graphed in Fig. 8(d).

The scaling parameter  $\gamma$  that governs an SCE is the ratio of the effective channel length and the natural channel length

$$\gamma = \frac{L_{\text{eff}}}{\lambda} \quad (4)^1$$

where  $L_{\text{eff}}$  is related to gate length, but as argued below, it is not generally the same.

Fig. 9(c) and (d) shows the change of  $S_{\text{min}}$  and DIBL versus gate length,  $L_g$ , normalized by scaling length,  $\lambda$ . The data all nearly fall on top of each other suggesting a universal electrostatic scaling trend.

MOSFET designs with better SCE immunity start suffering from SCE degradation at smaller  $\gamma$  values. In Si FDSOI MOSFETs, it is reported that the well-designed device architectures maintain acceptable SCE down to  $\gamma$  of 5–10 [29], [32]. Reports on InGaAs HEMTs indicate SCE immunity down to  $\gamma = 5$  [33]. In our device, we find that SCE FOMs start to degrade at  $L_g/\lambda \sim 15$ . This apparently early SCE degradation of our InGaAs MOSFETs might arise from two factors. First, here and in the analysis of III–V HEMTs [33], it is assumed that  $L_{\text{eff}} = L_g$ ; but HEMTs have an underlapped SD, and therefore the physical gate length  $L_g$  is an underestimation of the effective channel length  $L_{\text{eff}}$ , and hence  $\gamma$  (4), while the MOSFETs in this paper, as with modern very large-scale integration transistors, are designed with a slight  $n^+$ -gate overlap, and therefore  $L_g$  is close to or a slight overestimation of the effective channel length.

Second, in our device, the introduction of an n-type  $\delta$ -doped layer in the buffer below the channel is likely to degrade the subthreshold characteristics. This is equivalent to the introduction of positive fixed charge below the channel. In floating-body transistor designs, it has been observed that positive fixed charge beneath the channel results in degraded subthreshold characteristics in short III–V MOSFETs [3], [13] and FDSOI transistors [34]. This suggests that the appropriate transistor redesign might mitigate an SCE in future InGaAs QW-MOSFETs.

Our study reveals that planar InGaAs QW-MOSFETs, as fabricated here, even with the thinnest channels are at the limit of scaling at around  $L_g \sim 50 \text{ nm}$ . As in Si MOSFETs, the use of 3-D device architectures can increase  $\lambda$  and allows the device to scale to even smaller  $L_g$ . In any case, this will require the use of a very thin semiconductor body. This suggests that the tradeoff between OFF- and ON-state performance will remain an important engineering challenge [11], [29].

## V. CONCLUSION

We carry out a detailed vertical and lateral scaling study on InGaAs QW MOSFETs fabricated by a process that provides precise channel thickness control. We find that the intrinsic transconductance follows a classic gate length scaling behavior. The surface channel designs with a relatively thick

channel provide the best transconductance characteristics. The buried-channel designs feature lower transconductance presumably due to a loss of gate capacitance. Thin channels suffer from severe transconductance degradation. From a study of channel mobility, we identify the effective mass in the channel as a key factor affecting transport characteristics. In contrast with their excellent ON-state performance, the OFF-state characteristics of thick channel devices are significantly compromised. We find that the subthreshold swing and DIBL follow a classic scaling behavior similar to that of the Si FDSOI MOSFETs. The current InGaAs QW-MOSFET is at the limit of scaling around  $L_g \sim 50 \text{ nm}$ . This indicates the need for transistor redesign and advanced 3-D device architectures to deliver further progress.

## ACKNOWLEDGMENT

Device fabrication was carried out at the Microsystems Technology Laboratories and SEBL of MIT.

## REFERENCES

- [1] J. A. del Alamo *et al.*, “InGaAs MOSFETs for CMOS: Recent advances in process technology,” in *IEDM Tech. Dig.*, Dec. 2013, pp. 2.1.1–2.1.4.
- [2] M. J. W. Rodwell *et al.*, “Nanometer InP electron devices for VLSI and THz applications,” in *Proc. 72nd Annu. Device Res. Conf.*, Jun. 2014, pp. 215–216.
- [3] J. Lin, X. Zhao, T. Yu, D. A. Antoniadis, and J. A. del Alamo, “A new self-aligned quantum-well MOSFET architecture fabricated by a scalable tight-pitch process,” in *IEDM Tech. Dig.*, Dec. 2013, pp. 16.2.1–16.2.4.
- [4] J. Lin, D. A. Antoniadis, and J. A. del Alamo, “Sub-30 nm InAs quantum-well MOSFETs with self-aligned metal contacts and sub-1 nm EOT  $\text{HfO}_2$  insulator,” in *IEDM Tech. Dig.*, Dec. 2012, pp. 32.1.1–32.1.4.
- [5] A. Majumdar *et al.*, “CMOS-compatible self-aligned  $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$  MOSFETs with gate lengths down to 30 nm,” *IEEE Trans. Electron Devices*, vol. 61, no. 10, pp. 3399–3404, Oct. 2014.
- [6] H. J. Oh *et al.*, “Thermally robust phosphorous nitride interface passivation for InGaAs self-aligned gate-first n-MOSFET integrated with high- $k$  dielectric,” in *IEDM Tech. Dig.*, Dec. 2009, pp. 1–4.
- [7] S. Lee *et al.*, “High performance raised source/drain InAs/ $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$  channel metal-oxide-semiconductor field-effect transistors with reduced leakage using a vertical spacer,” *Appl. Phys. Lett.*, vol. 103, no. 3, p. 233503, 2013.
- [8] M. Egard *et al.*, “High transconductance self-aligned gate-last surface channel  $\text{In}_{0.53}\text{Ga}_{0.47}\text{As}$  MOSFET,” in *IEDM Tech. Dig.*, Dec. 2011, pp. 13.2.1–13.2.4.
- [9] S. W. Chang *et al.*, “InAs N-MOSFETs with record performance of  $I_{\text{on}} = 600 \mu\text{A}/\mu\text{m}$  at  $I_{\text{off}} = 100 \text{ nA}/\mu\text{m}$  ( $V_d = 0.5 \text{ V}$ ),” in *IEDM Tech. Dig.*, Dec. 2013, pp. 16.1.1–16.1.4.
- [10] S. Kim *et al.*, “Sub-60-nm extremely thin body  $\text{In}_x\text{Ga}_{1-x}\text{As}$ -on-insulator MOSFETs on Si with Ni-InGaAs metal S/D and MOS interface buffer engineering and its scalability,” *IEEE Trans. Electron Devices*, vol. 60, no. 8, pp. 2512–2517, Aug. 2013.
- [11] J. Lin, D. A. Antoniadis, and J. A. del Alamo, “Novel intrinsic and extrinsic engineering for high-performance high-density self-aligned InGaAs MOSFETs: Precise channel thickness control and sub-40-nm metal contacts,” in *IEDM Tech. Dig.*, Dec. 2014, pp. 25.1.1–25.1.4.
- [12] J. Lin, D. A. Antoniadis, and J. A. del Alamo, “Off-state leakage induced by band-to-band tunneling and floating-body bipolar effect in InGaAs quantum-well MOSFETs,” *IEEE Electron Device Lett.*, vol. 35, no. 12, pp. 1203–1205, Dec. 2014.
- [13] J. Lin, D. A. Antoniadis, and J. A. del Alamo, “Physics and mitigation of excess OFF-state current in InGaAs quantum-well MOSFETs,” *IEEE Trans. Electron Devices*, vol. 62, no. 5, pp. 1448–1455, May 2015.
- [14] S. Jin *et al.*, “Performance evaluation of InGaAs, Si, and Ge nFinFETs based on coupled 3D drift-diffusion/multisubband Boltzmann transport equations solver,” in *IEDM Tech. Dig.*, Dec. 2014, pp. 7.5.1–7.5.4.
- [15] J. Mo, E. Lind, G. Roll, and L.-E. Wernersson, “Reduction of off-state drain leakage in InGaAs-based metal-oxide-semiconductor field-effect transistors,” *Appl. Phys. Lett.*, vol. 105, no. 3, p. 033516, 2014.

<sup>1</sup>A factor of (1/2) is included in the original model. Here, it follows the definition without (1/2) for a similar study in [32] and [33].

- [16] A. Alian *et al.*, "Impact of the channel thickness on the performance of ultrathin InGaAs channel MOSFET devices," in *IEDM Tech. Dig.*, Dec. 2013, pp. 16.6.1–16.6.4.
- [17] J. Lin, D. A. Antoniadis, and J. A. del Alamo, "A CMOS-compatible fabrication process for scaled self-aligned InGaAs MOSFETs," in *Proc. CS MANTECH Conf.*, 2015, pp. 239–242.
- [18] J. Lin, T.-W. Kim, D. A. Antoniadis, and J. A. del Alamo, "A self-aligned InGaAs quantum-well metal-oxide-semiconductor field-effect transistor fabricated through a lift-off-free front-end process," *Appl. Phys. Exp.*, vol. 5, no. 6, p. 064002, 2012.
- [19] J. Lin, X. Zhao, D. A. Antoniadis, and J. A. del Alamo, "A novel digital etch technique for deeply scaled III–V MOSFETs," *IEEE Electron Device Lett.*, vol. 35, no. 4, pp. 440–442, Apr. 2014.
- [20] C.-H. Chang *et al.*, "Interfacial self-cleaning in atomic layer deposition of HfO<sub>2</sub> gate dielectric on In<sub>0.15</sub>Ga<sub>0.85</sub>As," *Appl. Phys. Lett.*, vol. 89, no. 24, p. 242911, 2006.
- [21] S. Deora *et al.*, "Positive bias instability and recovery in InGaAs channel nMOSFETs," *IEEE Trans. Device Mater. Rel.*, vol. 13, no. 4, pp. 507–514, Dec. 2013.
- [22] S. Y. Chou and D. A. Antoniadis, "Relationship between measured and intrinsic transconductances of FET's," *IEEE Trans. Electron Devices*, vol. 34, no. 2, pp. 448–450, Feb. 1987.
- [23] W. Lu, A. Guo, A. Vardi, and J. A. del Alamo, "A test structure to characterize nano-scale ohmic contacts in III–V MOSFETs," *IEEE Electron Device Lett.*, vol. 35, no. 2, pp. 178–180, Feb. 2014.
- [24] T. Chidambaram *et al.*, "Interface trap density and mobility extraction in InGaAs buried quantum well metal-oxide-semiconductor field-effect-transistors by gated Hall method," *Appl. Phys. Lett.*, vol. 104, no. 13, p. 131605, 2014.
- [25] N. Taoka *et al.*, "Impact of Fermi level pinning inside conduction band on electron mobility in InGaAs metal-oxide-semiconductor field-effect transistors," *Appl. Phys. Lett.*, vol. 103, no. 14, p. 143509, 2013.
- [26] D.-H. Kim *et al.*, "E-mode planar L<sub>g</sub> = 35 nm In<sub>0.7</sub>Ga<sub>0.3</sub>As MOSFETs with InP/Al<sub>2</sub>O<sub>3</sub>/HfO<sub>2</sub> (EOT = 0.8 nm) composite insulator," in *IEDM Tech. Dig.*, Dec. 2012, pp. 32.2.1–32.2.4.
- [27] Nextnano<sup>3</sup>. (2014). [Online]. Available: <http://www.nextnano.com/nextnano3/>
- [28] M. Levinshtein, S. Rumyantsev, and M. Shur, Eds., *Handbook Series on Semiconductor Parameters*, vol. 2. London, U.K.: World Scientific, 1999.
- [29] R.-H. Yan, A. Ourmazd, and K. F. Lee, "Scaling the Si MOSFET: From bulk to SOI to bulk," *IEEE Trans. Electron Devices*, vol. 39, no. 7, pp. 1704–1710, Jul. 1992.
- [30] M. Passlack *et al.*, "Dielectric properties of electron-beam deposited Ga<sub>2</sub>O<sub>3</sub> films," *Appl. Phys. Lett.*, vol. 64, no. 20, p. 2715, 1994.
- [31] S.-A. Lee, J.-Y. Hwang, J.-P. Kim, C.-R. Cho, W.-J. Lee, and S.-Y. Jeong, "Metal/insulator/semiconductor structure using Ga<sub>2</sub>O<sub>3</sub> layer by plasma enhanced atomic layer deposition," *J. Korean Phys. Soc.*, vol. 47, pp. S292–S295, Sep. 2005.
- [32] K. J. Kuhn, "Considerations for ultimate CMOS scaling," *IEEE Trans. Electron Devices*, vol. 59, no. 7, pp. 1813–1828, Jul. 2012.
- [33] D.-H. Kim and J. A. del Alamo, "Lateral and vertical scaling of In<sub>0.7</sub>Ga<sub>0.3</sub>As HEMTs for post-Si-CMOS logic applications," *IEEE Trans. Electron Devices*, vol. 55, no. 10, pp. 2546–2553, Oct. 2008.
- [34] F. E. Mamouni *et al.*, "Gate-length and drain-bias dependence of band-to-band tunneling-induced drain leakage in irradiated fully depleted SOI devices," *IEEE Trans. Nucl. Sci.*, vol. 55, no. 6, pp. 3259–3264, Dec. 2008.



**Jianqiang Lin** (S'08) received the B.Eng. (Hons.) and M.Eng. degrees in electrical engineering from the National University of Singapore, Singapore, in 2007 and 2009, respectively, and the Ph.D. degree from the Massachusetts Institute of Technology, Cambridge, MA, USA, in 2015.

He is currently a Post-Doctoral Associate with the Massachusetts Institute of Technology.



**Dimitri A. Antoniadis** (M'79–SM'83–F'90–LF'14) received the B.S. degree in physics from the National University of Athens, Athens, Greece, in 1970, and the M.S. and Ph.D. degrees in electrical engineering from Stanford University, Stanford, CA, USA, in 1972 and 1976, respectively.

He joined the Massachusetts Institute of Technology, Cambridge, MA, USA, in 1978, where he is currently the Ray and Maria Stata Professor of Electrical Engineering.



**Jesús A. del Alamo** (S'79–M'85–SM'92–F'06) received the Telecommunications Engineer degree from the Polytechnic University of Madrid, Madrid, Spain, in 1980, and the M.S. and Ph.D. degrees in electrical engineering from Stanford University, Stanford, CA, USA, in 1983 and 1985, respectively.

He has been with the Massachusetts Institute of Technology, Cambridge, MA, USA, since 1988, where he is currently a Donner Professor and a MacVicar Faculty Fellow.